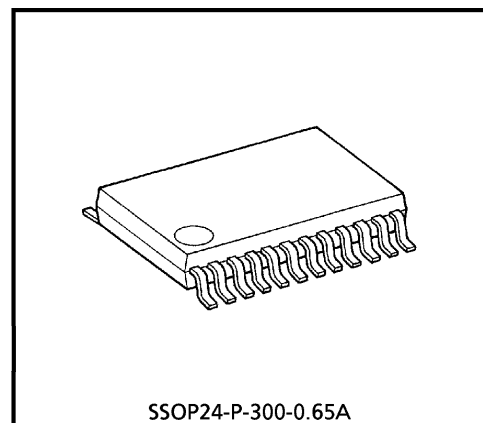


T A 3 1 1 3 7 F N G

IF DETECTOR IC FOR CORDLESS AND CELLULAR PHONES

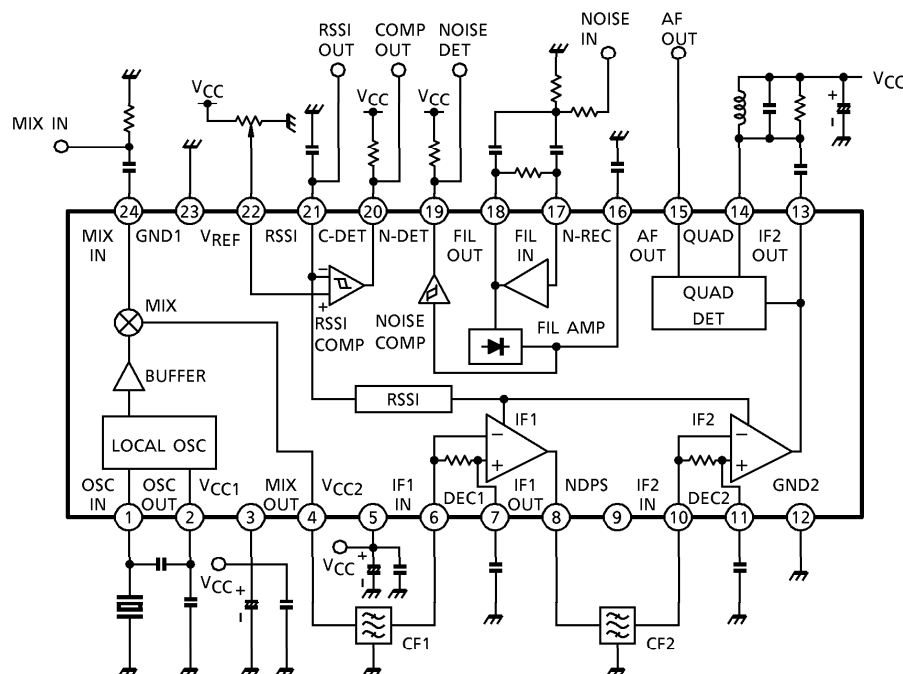
FEATURES

- Low operating voltage : $V_{CC} = 1.8 \sim 6.0V$
- Excellent temperature characteristic
- High selectivity can be designed by 2 IF amps
- High sensitivity
12dB sensitivity : $8.5dB\mu V$ EMF (50Ω)
- Intercept point is very high : $107dB\mu V$ (0dBm)
- Built-in 2nd MIX
Operating frequency : $10 \sim 150MHz$
- Built-in noise detection circuit
- RSSI function
- RSSI COMPARATOR
- Small current consumption : $I_{CC} = 3.8mA$ (Typ.)
- Small package : SSOP 24 pin (0.65mm pitch)



Weight : 0.14g (Typ.)

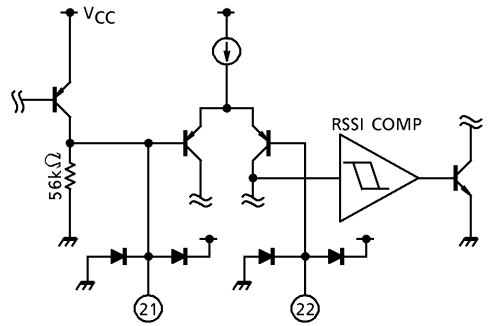
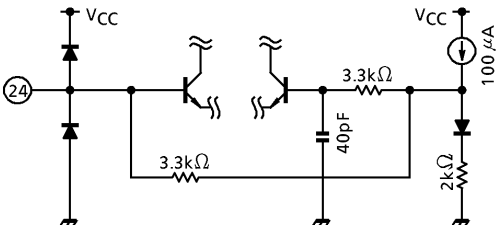
BLOCK DIAGRAM



PIN FUNCTION (The values of resistor and capacitor are typical.)

| PIN No. | PIN NAME | FUNCTION | INTERNAL EQUIVALENT CIRCUIT |
|---------|----------|---|-----------------------------|
| 1 | OSC IN | LOCAL OSC input and output terminals. Colpitts oscillator is formed by internal emitter follower and external X'tal. And external injection is possible from pin 2. | |
| 2 | OSC OUT | | |
| 3 | VCC1 | Power supply 1 | — |
| 4 | MIX OUT | MIX Output terminal. Output impedance is around 1.8kΩ. | |
| 5 | VCC2 | Power supply 2 | — |
| 6 | IF1 IN | IF1 input and decoupling for bias. Input impedance is around 1.8kΩ. | |
| 7 | DEC1 | | |
| 8 | IF1 OUT | Output terminal of IF1 AMP. | |
| 9 | NDPS | Connect to VCC | — |
| 10 | IF2 IN | IF2 input and decoupling for bias. Input impedance is around 1.8kΩ. | |
| 11 | DEC2 | | |
| 12 | GND2 | GND terminal | — |
| 13 | IF2 OUT | Output terminal of IF2 AMP. | |

| PIN No. | PIN NAME | FUNCTION | INTERNAL EQUIVALENT CIRCUIT |
|---------|----------|--|-----------------------------|
| 14 | QUAD | Phase input terminal of FM DETECTOR. | |
| 15 | AF OUT | Demodulate signal output terminal. Carrier leak is small as LPF is built-in. Output impedance is around 360Ω. | |
| 16 | N-REC | After output of inverter amp amplified around 20dB, noise signal is rectified by external capacitor. | |
| 17 | FIL IN | Inverter amp input and output terminal. BPF is composed of external capacitors and resistors. Connected internally to rectifier circuit by coupling capacitor. | |
| 18 | FIL OUT | | |
| 19 | N-DET | The result of noise detection is output by comparing output voltage of N-REC terminal with internal reference. Hysteresis range is about 100mV and output is open collector. | |
| 20 | C-DET | Comparison output terminal of VREF terminal input voltage and RSSI terminal output voltage. When VREF < RSSI, C-DET output is "L" level. | |

| PIN No. | PIN NAME | FUNCTION | INTERNAL EQUIVALENT CIRCUIT |
|---------|----------|--|--|
| 21 | RSSI | This terminal outputs DC level according to input signal level to IF AMP. Dynamic range is around 70dB. |  |
| 22 | VREF | Reference voltage input terminal. | |
| 23 | GND1 | GND terminal | — |
| 24 | MIX IN | 1st IF signal input terminal. |  |

1. LOCAL OSC external injection method

Inject as shown in Fig. 1. Setting the injection level between $95\text{dB}\mu\text{V}$ and $100\text{dB}\mu\text{V}$. A built-in buffer amp minimizes leakage from the MIX.

Input from pin 1 is possible as shown in Fig. 2. However, when the input frequency is high, the level at pin 2 may not be sufficient, causing a decrease in sensitivity. In such a case, add resistor R_{51} and set the input signal so that signal level at pin 2 is adequate. The input capacitance of pins 1 and 2 are respectively 2.4pF (Typ.) and 4.5pF (Typ.).

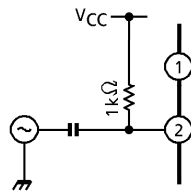


Fig. 1

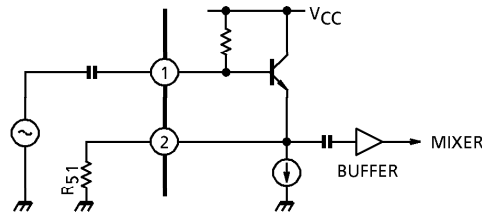


Fig. 2

2. Overtone oscillation

Fig. 3 shows the basic configuration of the local oscillation circuit using overtone oscillation. The C_{51} and L_1 tuning circuits prevent crystal fundamental oscillation. Therefore, set C_{51} and L_1 to inductive at the fundamental frequency and capacitive at the overtone frequency. Since the level at pin 2 may decrease and the sensitivity may fall at high frequency as with external injection, adjust the oscillation level using R_{51} .

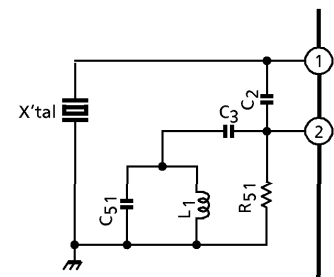


Fig. 3

3. Detection circuit

Demodulation output can be increased by raising damping resistance R_3 . However, be careful because the temperature dependency of the modulation output also increases. The demodulation output depends largely on C_{101} . For C_{101} , use a capacitor with good temperature characteristics.

4. NDPS terminal

The NDPS terminal (pin 9) is used for the power source for the 16-19 pin block. When pins 16-19 are not in use, current consumption can be reduced by opening the NDPS terminal. (In this case, pins 16-19 can be open.)

5. C-DET terminal

The C-DET terminal (pin 20) is used for the comparator output of the V_{REF} terminal (pin 22) output voltage and the RSSI terminal (pin 21) output voltage.

When $V_{REF} > RSSI$, C-DET = "H"

When $V_{REF} < RSSI$, C-DET = "L"

* The hysteresis range is about 30mV (Typ.).

When not in use, set pin 20 to open and connect pin 22 to V_{CC} .

6. Inverter amp usage

The inverter amp can be used to form a band pass filter as shown in Fig. 4.

Set constants as in equations (1) to (3). However, because a low pass filter and a high pass filter are built in, it is recommended that center frequency f_0 be about 30kHz.

$$(1) \quad f_0 = \frac{1}{2\pi\sqrt{R_4(R_5 // R_6)C^2}}$$

$$(2) \quad G_V = R_4 / 2 \times R_5$$

$$(3) \quad Q^2 = \frac{R_4}{4(R_5 // R_6)}$$

at $R_5 \gg R_p$

Example: $R_4 = 150k\Omega$, $R_5 = 330k\Omega$

$R_6 = 3.3k\Omega$, $R_p = 20k\Omega$ (VR)

When $C = 220pF$

$f_0 \approx 31kHz$, $G_V \approx -13dB$

$Q^2 \approx 12$

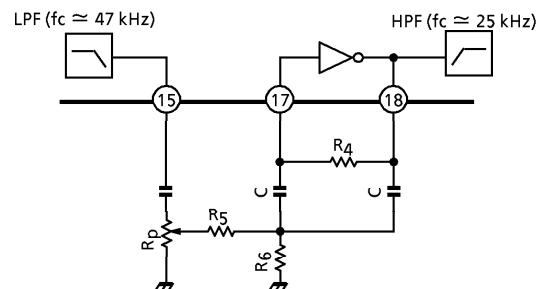


Fig. 4

7. Noise detection rise time

The rise time is a proportion of time constant 7.5ms of the smoothing capacitor $C_g = 0.1\mu F$ of the noise rectifier and internal resistor $75k\Omega$. Although decreasing the capacitance of C_g can shorten the rise time, note that the noise detection output fluctuation may increase. This should be taken into account before use.

8. RSSI function

A DC voltage corresponding to the input level of IF input pins (pin 6 and pin 10) is output to the RSSI pin (pin 21). While the linear range is about 70dB when $V_{CC}=2V$, the range can be expanded to 80dB as in Fig. 5.

However, in such a case, note that the temperature characteristics of the RSSI output may alter due to a disparity between the temperature coefficient of the external resistor and the internal resistance of the IC.

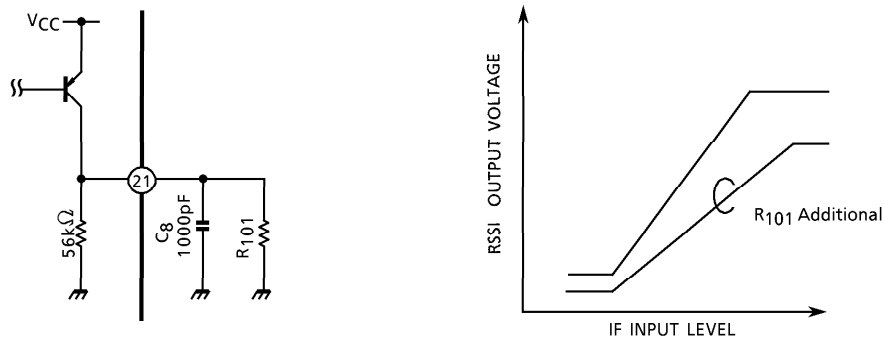


Fig. 5

9. DC voltages for pins (Typical values for reference)

$V_{CC}=3.0V$

| PIN No. | PIN NAME | VOLTAGE | PIN No. | PIN NAME | VOLTAGE |
|---------|-----------|---------|---------|----------|---------|
| 1 | OCS IN | 2.98 | 13 | IF2 OUT | 2.14 |
| 2 | OSC OUT | 2.28 | 14 | QUAD | 3.0 |
| 3 | V_{CC1} | 3.0 | 15 | AF OUT | — |
| 4 | MIX OUT | 1.76 | 16 | N-REC | — |
| 5 | V_{CC2} | 3.0 | 17 | FIL IN | 0.64 |
| 6 | IF1 IN | 2.58 | 18 | FIL OUT | 0.66 |
| 7 | DEC 1 | 2.58 | 19 | N-DET | — |
| 8 | IF1 OUT | 2.0 | 20 | C-DET | — |
| 9 | NDPS | 3.0 | 21 | RSSI | — |
| 10 | IF2 IN | 2.65 | 22 | VREF | — |
| 11 | DEC 2 | 2.65 | 23 | GND 1 | 0.0 |
| 12 | GND 2 | 0.0 | 24 | MIX IN | 0.93 |

(Unit : V)

MAXIMUM RATINGS (Ta = 25°C)

| CHARACTERISTIC | SYMBOL | RATING | UNIT |
|-----------------------|------------------|----------|------|
| Supply Voltage | V _{CC} | 7 | V |
| Power Dissipation | P _D | 780 | mW |
| Operating Temperature | T _{opr} | − 30~85 | °C |
| Storage Temperature | T _{stg} | − 50~150 | °C |

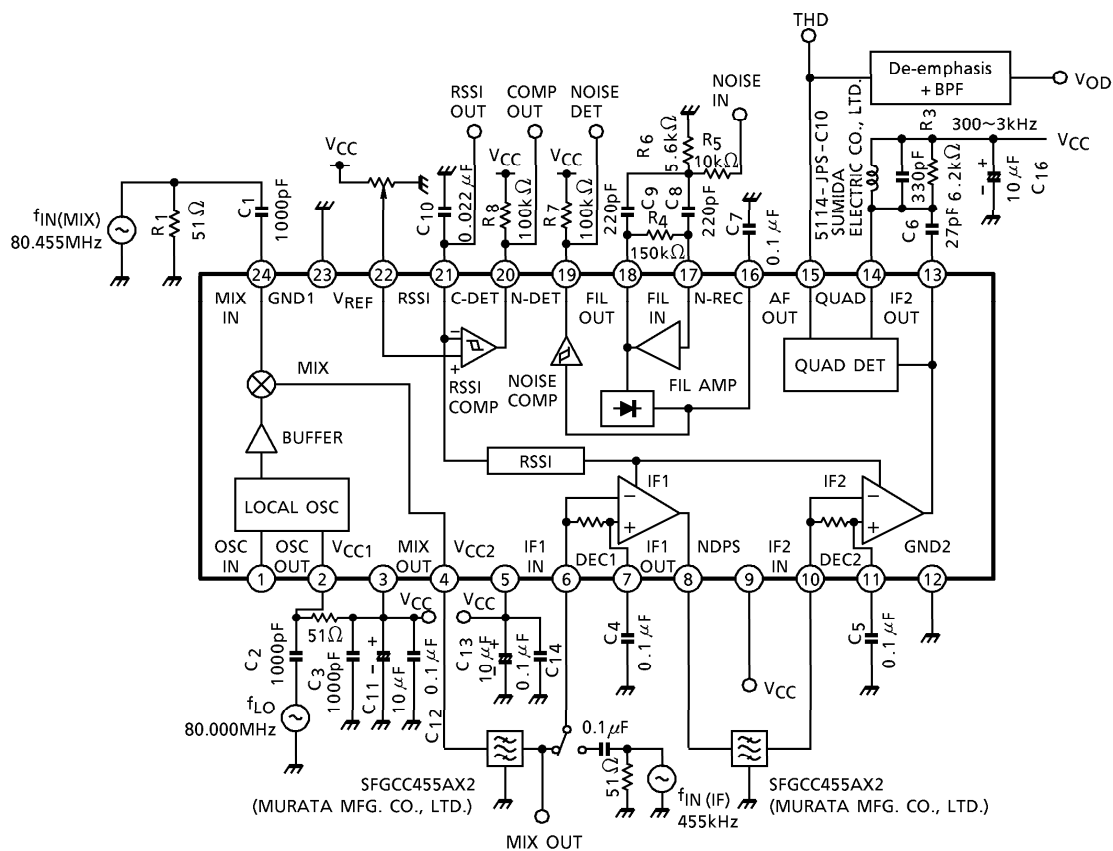
ELECTRICAL CHARACTERISTICS

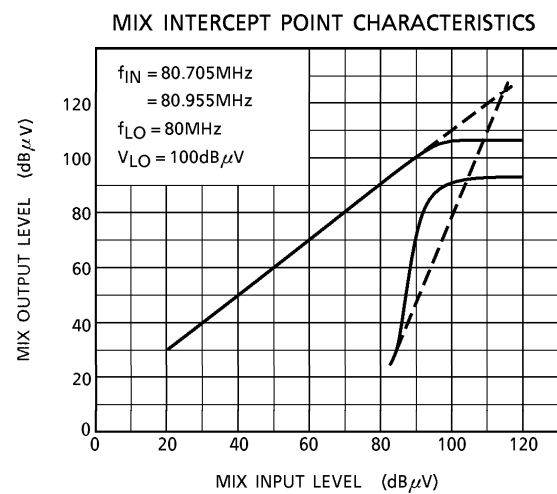
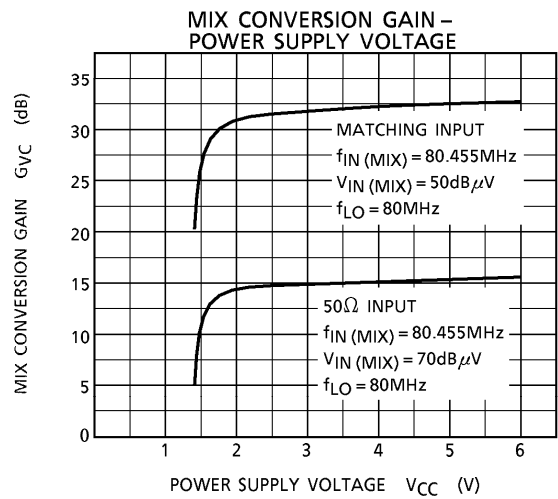
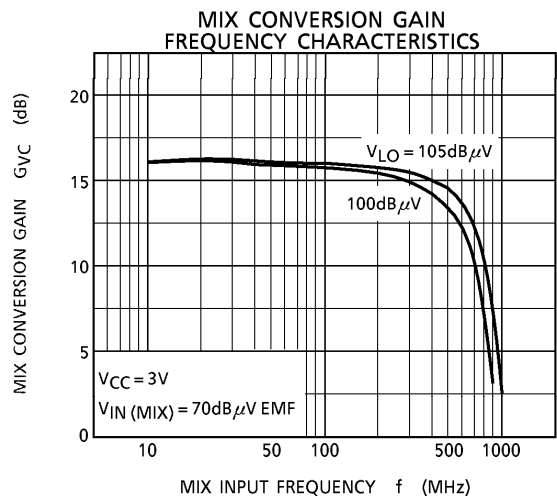
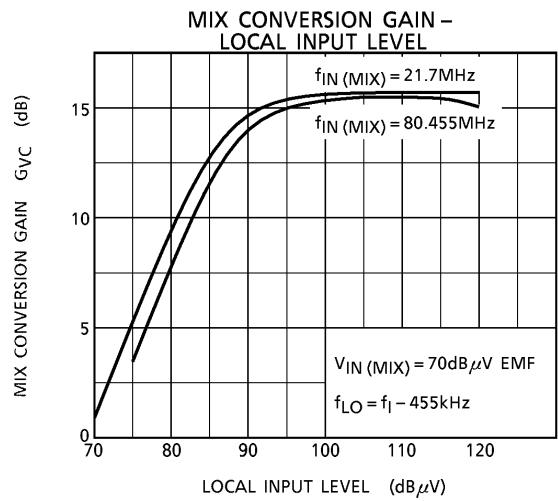
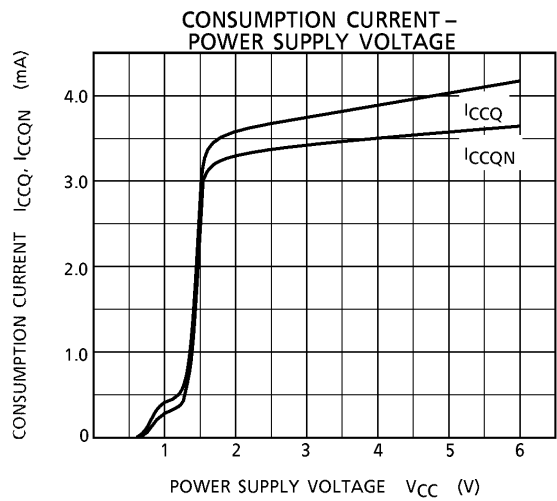
(Unless otherwise specified, V_{CC} = 3.0V, f_{IN} (MIX) = 80.455MHz, Δf = ± 8.0kHz, f_{MOD} = 1kHz, Ta = 25°C, f_{LO} = 80MHz, V_{LO} = 100dBμV, f_{IN} (IF) = 455kHz, V_{IN} (IF) = 60dBμV EMF)

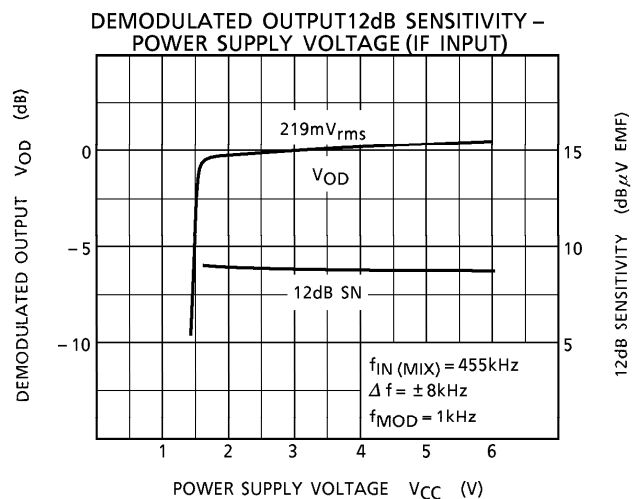
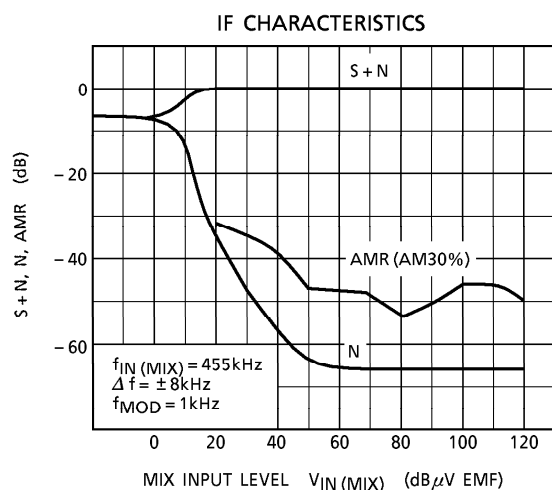
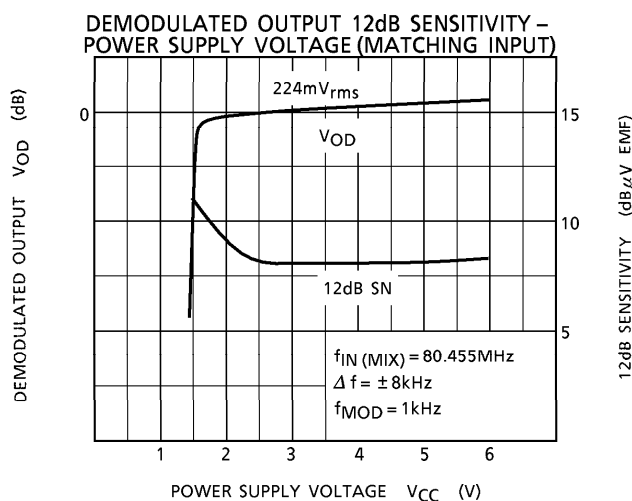
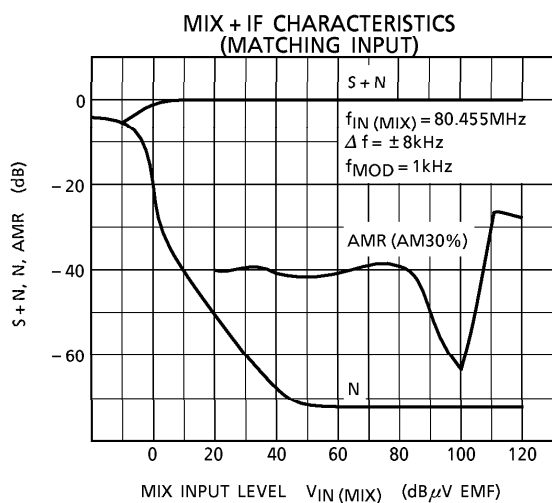
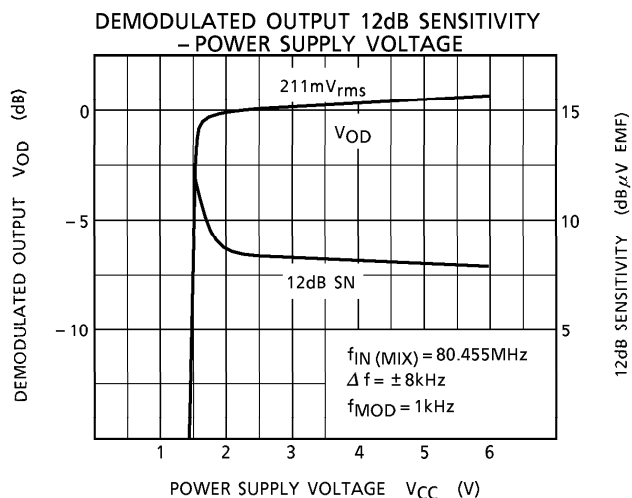
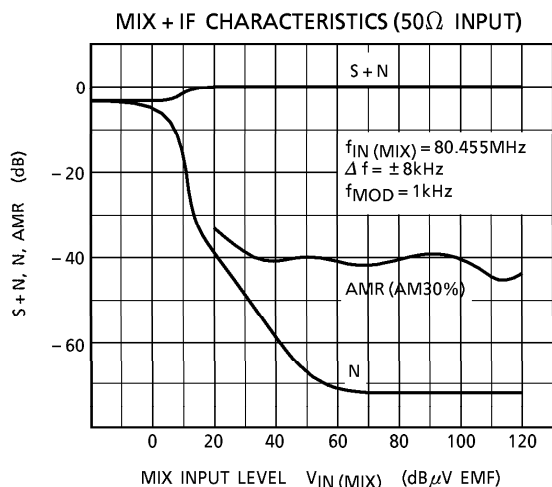
| CHARACTERISTIC | SYMBOL | TEST CIRCUIT | TEST CONDITION | MIN. | TYP. | MAX. | UNIT |
|-----------------------|------------------------|--------------|---|------|------|------|-------------------|
| Supply Voltage | V _{CC} | — | — | 1.8 | 3.0 | 6.0 | V |
| Quiescent Current | I _{CCQ} | — | With NOISE CIRCUIT | — | 4.1 | 5.6 | mA |
| | I _{CCQN} | — | Without NOISE CIRCUIT | — | 3.8 | 5.2 | mA |
| MIX Conversion Gain | G _{VC} | — | Measurement after ceramic filter | 12.5 | 15.5 | 18.5 | dB |
| MIX Intercept Point | P _{IM} | — | — | — | 107 | — | dBμV |
| MIX Input Impedance | R _{IN} (MIX) | — | — | — | 4.5 | — | kΩ |
| | C _{IN} (MIX) | — | — | — | 2.4 | — | pF |
| MIX Output Resistance | R _{OUT} (MIX) | — | — | 1.2 | 1.8 | 2.4 | kΩ |
| 12dB Sensitivity | 12dB SN | — | 50Ω Input | — | 8.5 | — | dBμV EMF |
| Demodulated Output | V _{OD} | — | V _{IN} (IF) = 60dBμV EMF | 160 | 210 | 280 | mV _{rms} |
| SN Ratio | SN | — | V _{IN} (IF) = 60dBμV EMF | 50 | 65 | — | dB |
| AM Rejection Ratio | AMR | — | V _{IN} (IF) = 60dBμV EMF, AM = 30% | — | 48 | — | dB |
| IF1 Gain | G (IF1) | — | — | — | 25 | — | dB |
| IF2 Gain | G (IF2) | — | — | — | 77 | — | dB |
| IF1 Input Resistance | R _{IN} (IF1) | — | — | 1.2 | 1.8 | 2.4 | kΩ |
| IF1 Output Resistance | R _{OUT} (IF1) | — | — | 1.2 | 1.8 | 2.4 | kΩ |
| IF2 Input Resistance | R _{IN} (IF2) | — | — | 1.2 | 1.8 | 2.4 | kΩ |
| AF Output Impedance | Z _{AF} | — | — | — | 360 | — | Ω |
| RSSI Output Voltage | V _{RSSI-1} | — | V _{CC} = 3V, V _{IN} (IF) = 20dBμV EMF | 0.1 | 0.25 | 0.45 | V |
| | V _{RSSI-2} | — | V _{CC} = 3V, V _{IN} (IF) = 100dBμV EMF | 1.8 | 2.2 | 2.6 | V |

| CHARACTERISTIC | SYMBOL | TEST CIR- CUIT | TEST CONDITION | MIN. | TYP. | MAX. | UNIT |
|--|------------------|-------------------|---|------|------|--------------|---------------|
| Noise Detection Output Voltage | V_{NDET} | — | $I_{SINK} = 0.2\text{mA}$ | — | 0.1 | 0.5 | V |
| Noise Detection Out Leak Current | I_{LEAK} | — | $V_{NREC} = 0.6\text{V}$, $V_{NDET} = 2\text{V}$ | — | 0 | 5 | μA |
| Noise Detection Level | "H" Level | V_{TH-H} | — | — | 0.5 | 0.7 | V |
| | "L" Level | V_{TH-L} | — | 0.3 | 0.4 | — | V |
| RSSI COMPARATOR Range Of Hysteresis | V_{HYS} | — | — | — | 30 | — | mV |
| RSSI COMPARATOR Input Range Of Reference | ΔV_{REF} | — | — | 0.3 | — | $V_{CC} - 1$ | V |

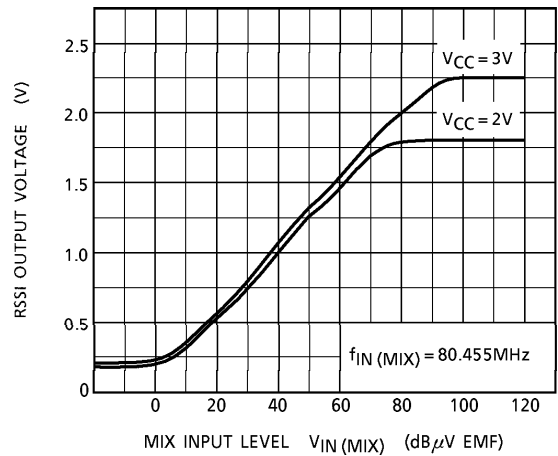
TEST CIRCUIT



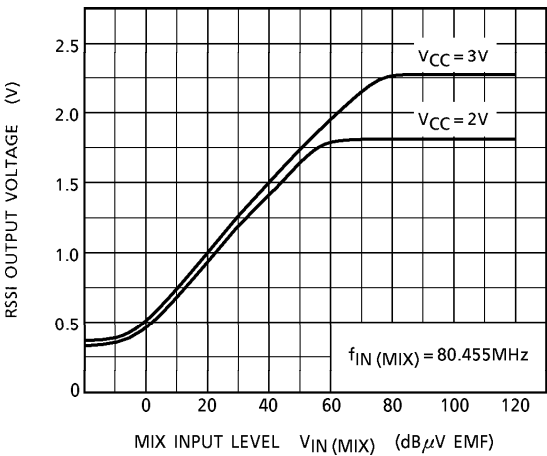




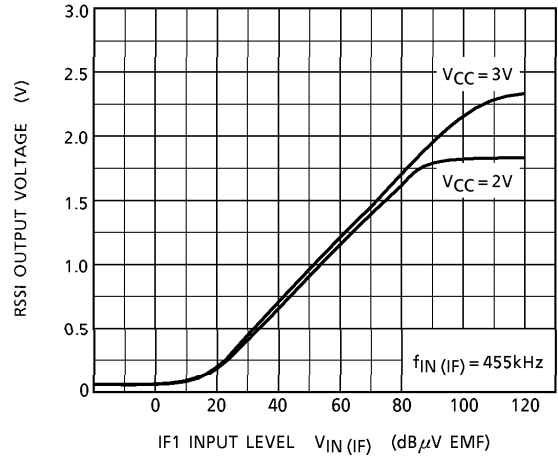
RSSI CHARACTERISTICS (50Ω INPUT)



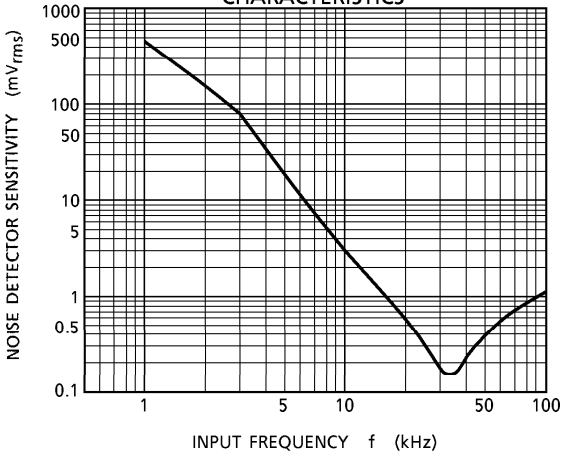
RSSI CHARACTERISTICS (MATCHING INPUT)



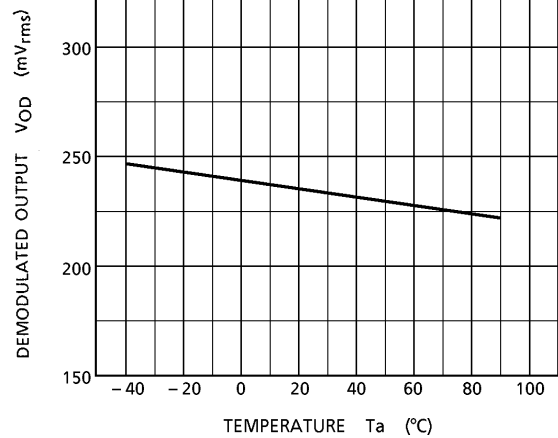
RSSI CHARACTERISTICS (IF1 INPUT)



NOISE DETECTION FREQUENCY CHARACTERISTICS



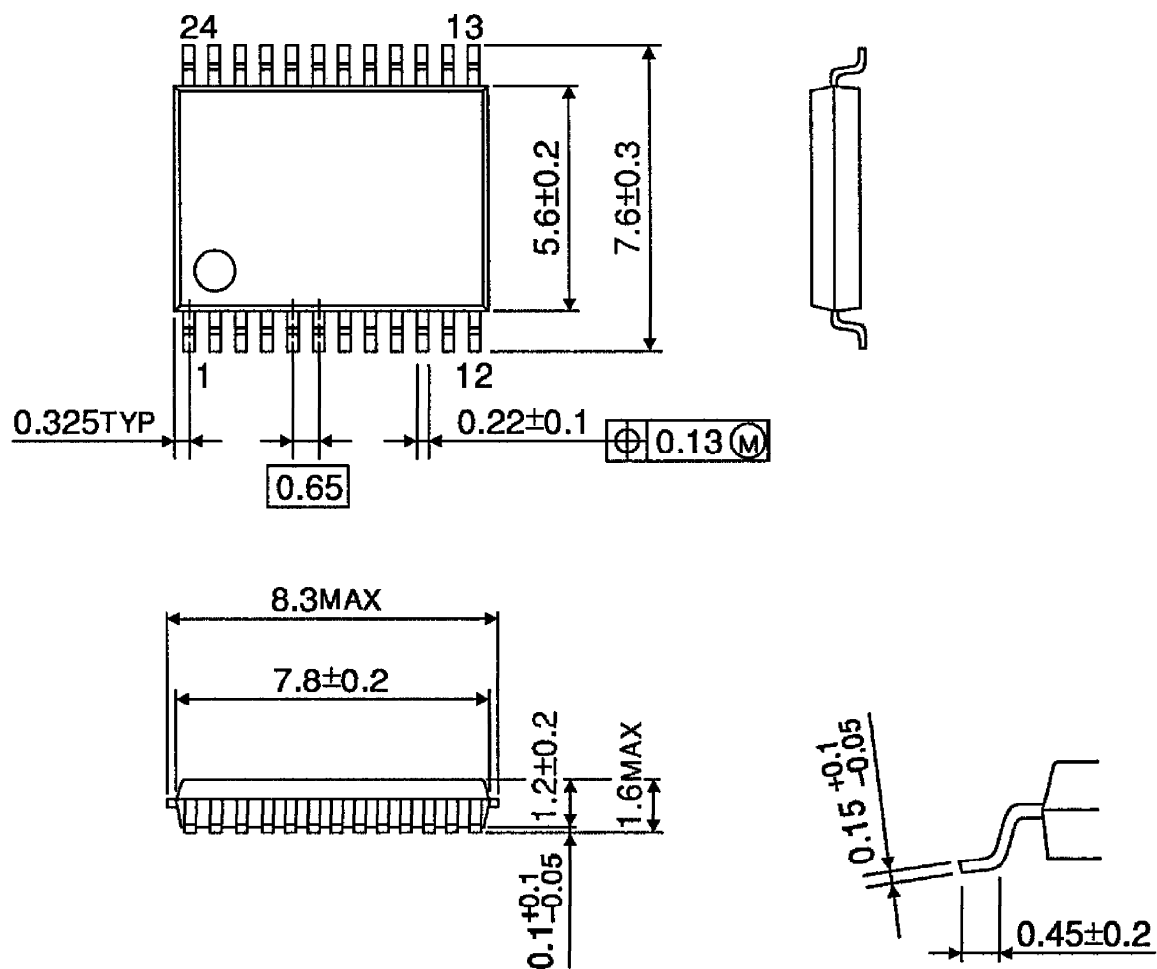
DEMODULATED OUTPUT TEMPERATURE CHARACTERISTICS



The schematic diagram illustrates the internal architecture of the SFGCC455AX2. Key components include a TCXO (80.000 MHz) connected to pins 1 and 2, a LOCAL OSC block, a BUFFER, a MIX block, and a QUAD DET block. The diagram also shows various capacitors (C1 through C16) and resistors (R1 through R9) used for tuning and signal processing. The output of the QUAD DET is connected to a De-emphasis + BPF block, which produces the final V_{OD} output. The diagram is labeled with SFGCC455AX2 (MURATA MFG. CO., LTD.) at the bottom.

SSOP24-P-300-0.65A

Unit : mm



Weight : 0.14g (Typ.)

About solderability, following conditions were confirmed

- Solderability
 - (1) Use of Sn-63Pb solder Bath
 - solder bath temperature = 230°C
 - dipping time = 5 seconds
 - the number of times = once
 - use of R-type flux
 - (2) Use of Sn-3.0Ag-0.5Cu solder Bath
 - solder bath temperature = 245°C
 - dipping time = 5 seconds
 - the number of times = once
 - use of R-type flux

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030619EBA

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